DISPERSION COMPENSATION USING ALL PASS FILTERS IN OPTICAL FIBERS

Shivinder Devra¹ and Gurmeet Kaur²

¹Student University College of Engineering, Punjabi University, Patiala, India ²Associate Professor University College of Engineering, Punjabi University, Patiala, India E-mail:¹shivinderdevra@yahoo.co.in, ²farishta02@yahoo.co.in

Abstract: All pass filters (APFs) are used in dispersion compensation which is the foremost requirement in an optical fiber link. All pass filters can correct any order of dispersion by the careful design of multistage all pass filters starting from very simple components with the use of N port devices. Multiple channels, as in wavelength division multiplexed (WDM) system, can be compensated with a single device since these filters are periodic in phase response. The design technique and implementation of these filters has been discussed in this paper.

Keywords: Optical communication, optical fibers, wavelength division multiplexed systems, dispersion compensation all pass filters.

1. Introduction

All Pass filters are used to compensate the chromatic dispersion in wavelength division multiplexed (WDM) optical fiber communication system [1]. Optical fiber communication is a way of transmitting the information from one place to another by modulating the light signal with the information signal. The light signal required for communication is generated using the spontaneous and stimulated emission occurring in light emitting diodes (LEDs) and LASERs [2]. Since the energy levels are not discrete so mono-chromaticity of the light signal is lost and it introduces chromatic dispersion. The number of compensating techniques has been reported in the literature [3], [4], [5], [6] including dispersion compensating fibers (DCFs), Fiber Bragg gratings (FBGs), Electronic Dispersion compensation (EDC) each having its own advantages and disadvantages. In WDM system where a number of frequencies are interleaved, dispersion is compensated using all pass filters [7]. All pass filters are linear systems having variable phase response and constant amplitude response. The variable phase response of the APFs makes them to be used as the phase equalizers to compensate the chromatic dispersion.

The need of dispersion compensation, general properties of all pass filters, the design and implementation of all pass filters along with tunable dispersion compensation all pass filters have been discussed in this paper.

2. Need

Due to the presence of chromatic dispersion the light pulse carrying the required information is spread into various components and each component travel differently along the optical fiber with different velocity and hence reach at the receiver at different times which distorts the information and can't be interpreted in the correct manner This is called group velocity dispersion (GVD) which cause the light pulses to spread in fibers, degrading signals over long distances [8-11]. In order to remove the spreading of the optical or light pulses, the dispersion compensation is the most key feature required in optical fiber communication system.

The traditional techniques like DCFs, FBGs, and EDC are not suitable for dispersion compensation in WDM system. DCFs give high insertion loss, large footprint, and non-linear distortions when the input signal is high etc. Also for the multiple channels in WDM system, the number of DCFs has to be installed making the system complex and costly. The same problem is with the FBGs which compensate the dispersion by the recompression of an optical signal. For different frequencies different architectures of the

FBGs have to be introduced along the fiber link. EDC is rendered ineffective for WDM system since it is complex and also not a direct method of compensation as it involve the optical to electronic and electrical to optical conversions making the WDM communication slow which can't be tolerated in this growing world hence the need of all pass filters is realized by which the multiple channels can be compensated with a single device because of the periodic properties of the phase response of these filters [12-15].

3. All Pass Filters (APF)

The dispersion compensation using digital filters is a new technique for the removal of phase distortions of an optical signal. After the various channels have been multiplexed by the wavelength interleaver over the single fiber the next step is to compensate the phase distortions due to different group delays for different channels [1], [7]. Dispersion compensating fibers [2], [3] (with opposite chromatic dispersion as that of channels) are not used these days as they introduce large footprint, high insertion loss, introduce nonlinear distortion etc, hence they have been replaced by all pass filter structure. It is a special filter with flat magnitude spectrum and non-linear phase spectrum, so it compensates phase distortion without affecting magnitude spectrum of signals [8], [9]. These all pass filters (APF's) are linear systems, which have an amplitude response that is constant over all frequencies and a phase response that varies with frequency. The period of frequency response of all pass filters is usually referred to as free spectral range (FSR). Mathematically, the frequency response of a filter is written as

$$H(\omega) = |H(\omega)| \exp[j\phi(\omega)]$$
(1)

then for an APF $|H(\omega)| = c$ where c is a constant and $\phi(\omega)$ can be made arbitrarily close to any desired phase response. With this characteristic the nth-order dispersion is evaluated as $1/FSR^n$, further group delay can be enhanced by adding more number of stages [12]. However it increases loss in the system. Adding stages to the APF help in recovering group delay that is lost when the FSR is increased [12], [13].

The dispersion compensation obtained experimentally is

$$D \sim N/FSR^2 \Delta^2$$
 (2)

Where N is number of channels and Δ is distance of poles and zeros of the unit circle. The dispersion may be increased by reducing the FSR with the introduction of more number of stages or by reducing the Δ_{\perp}

4. APF Design and Implementation

For the design of an APF, a four port device with frequency independent matrix elements can be considered. By connecting any one of the outputs through a delay to any one of the inputs a single stage APF can be realized [12]. APF may be implemented using Directional couplers, Mach-Zehnder interferometer, and thin film filter as shown below in Fig.1-4:



Fig 2. Two stage APF [12].



Fig 3.Single stage APF using Mach-Zehnder interferometer [12]



Fig 4. Thin-film example: (a) Interface between two dielectrics. The scattering matrix relates the "input" amplitudes A0 and B1 to the "output" amplitudes A1 and B0. (b) By connecting the "output" A1 to the "input" B1 through a delay (using a 100% reflector a distance d away), a single-stage APF is obtained. This is exactly the familiar Gires–Tournois interferometer [12].

5. Tunable Dispersion Compensation All Pass Filters

Chromatic dispersion compensation is critical for high bit rate light wave systems. Reconfigurable optical networks introduce a need for tunable dispersion compensation since different routes may have different cumulative dispersions [14]. In addition, tunable dispersion compensation is required for high bit rate nonlinear systems whose optimal dispersion depends on the channel power which may fluctuate over time. Different wavelengths have different cumulative dispersions at the receiver, and a device capable of applying varying amounts of dispersion compensation to each channel is needed. Because of the large number of channels in dense wavelength-division-multiplexed (WDM) systems, periodic filters are advantageous compared to single channel devices which require a unique filter for every WDM channel [14], [15]. Tunable dispersion compensation filters are of two types:

5.1 MEMS Compensation All Pass Filters

The tunable all-pass filter is based on the mechanical antireflection switch (MARS) device, which is a variable- thickness Fabry–Perot cavity consisting of a silicon substrate, an air gap, and a quarter-wave thick dielectric membrane. A silicon nitride layer is used for the membrane, and the gap is nominally $3 \lambda_0/4$. The cavity formed by the membrane and top surface of the substrate yields a reflection of about 70%. The gap is varied from $3 \lambda_0/4$ to $\lambda_0/4$ by applying a voltage to electrodes on top of the membrane as shown in Fig.5.



Fig 5. MEMS all-pass filter schematic showing the change in air gap with applied voltage [14].

The voltage creates an electrostatic force that pulls the membrane closer to the substrate surface, while the membrane tension provides a linear restoring force. At a gap of $\lambda_0/4$, the reflection is reduced to ~ 0% since the silicon nitride acts as an antireflection coating for the silicon substrate. To make an all-pass filter, the aim is to use Fabry–Perot cavity as a tunable, partial reflector and add a high reflectance coating to the

back side of the substrate [14]. A reflectivity > 97% is obtained using a multi-layer stack. The substrate thickness L determines the free spectral range FSR = c / $2n_gL$, where n_g is the group index. For a 100-GHz FSR, the silicon thickness is 411 m. By selecting the filter period equal the channel spacing in a WDM system, multiple channels can be compensated. The filter dispersion is D = d τ /d λ (ps/nm). For a completely tunable all-pass filter, both the partial reflector and the cavity optical length must be tunable. By varying the applied voltage, the partial reflectance of the front mirror is changed. For tuning Φ_n , the substrate is mounted on a thermo-electric cooler, and the cavity optical thickness is tuned via the thermo-optic effect. Tuning of the cavity length can also be used to compensate for variations in the fabricated cavity length from the design nominal [14].

5.2 Integrated All Pass Filters for Tunable Dispersion Compensation

Two parameters control its group delay response, the phase Φ and power coupling ratio k_r. By using a multistage filter where the parameters are chosen optimally for each stage, a constant dispersion (or any desired response) can be approximated over a large portion of the FSR, thus yielding a large bandwidth utilization factor [15]. It is critical to achieve the design values for these parameters, and fabrication-induced variations on the coupling ratios must be minimized. The new all-pass filter architectures are shown in Fig.6 (a), (b) and (c).



Fig 6. (a) Ring resonator all-pass filter with a fixed coupling ratio, and fully tunable ring resonator all-pass filters with (b) an asymmetric MZI and (c) a symmetric MZI [15].

The single coupler is replaced with a Mach–Zehnder interferometer (MZI). The MZI is curved to minimize any increase in the feedback path length. The advantage is that a phase shifter can be used to tune the effective coupling k_e into the feedback path, thus a completely tunable all-pass filter is easily realized with two phase shifters, one to set k_e and one to tune the resonant wavelength. The tolerances on the couplers k composing the MZI are substantially relaxed compared to the tolerance on k_r . In Fig. (b), the MZI path lengths are different by a length $\Delta L = \pi d_{sep}$ where d_{sep} is the separation of the MZI arms. The effective coupling is given by k_e which can be set to zero at a given wavelength by choosing Φ_m appropriately [15].

$$k_{e} = 4k(1 - k)\cos^{2}(\left[2\pi n_{g} \Delta L/\lambda + \Phi_{m}\right]/2)$$
(3)

)

In Fig. (c),

$$k_e = 1 - 4 k(1 - k) \cos^2 (\Phi_m/2)$$
Hence $k_e = 1$ can be achieved by the proper choice of Φ_m . (4)

6. Conclusion

There are number of techniques to compensate the chromatic dispersion of an optical signal travelling along the optical fiber. The dispersion compensation using digital filters is the most effective way of compensating it. It is a new class of digital filters implemented in the optical domain called all pass filters. All pass filters are lossless filters which offer the flexibility to tune a desired phase response arbitrarily close by increasing the number of stages keeping magnitude response of a system unchanged. The fully tunable all pass filters having 100 GHz FSR and negligible polarization dependence have been fabricated with tuning range of \pm 100 ps/nm, a pass band width of 50 GHz and group delay ripple of <3-ps peak are demonstrated. With the careful design of APF's together with the feedback equalization used at the receiver, the 10Gbps WDM system with FSR = 50GHz, OSNR = 22.7 at BER of 10⁻⁹ may be realized.

7. References

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